

# Hi-Rel DC/DC CONVERTER MGDD-80: 80W POWER



# 8:1 Ultra Wide Input Dual Outputs Metallic Case - 1 500 VDC Isolation

- Ultra wide input range 9-60 VDC
- . Nominal power up to 80 W
- Maximum power per channel up to 70W
- Nominal dual output voltage from 5V to 24V
- High efficiency over the entire range (typ. 92%)
- Soft start
- Galvanic isolation 1.500 VDC
- Integrated LC input filter
- Permanent short circuit protection
- External synchronisation
- External trim adjustment : -20/+10%
- No optocoupler for high reliability
- OTP-Disable (OTP-D) function
- RoHS process



The MGDD-80 ultra wide input series designates a full family of DC/DC power modules with a permanent ultra wide input voltage range of 9-60 volts. The family is designed for use in distributed power architecture where variable input voltage and transient are prevalent making them ideal particularly for avionics and military applications. The MGDD-80 is ideal for applications where high power density up to 50W/inch³ is required.

The MGDD-80 series is compliant with D0-160 and MIL-STD-704 transient voltage without additional voltage limiter.

The serie includes dual output voltage choices individually isolated of 2 x 3.3 volts, 2 x 5 volts , 2 x 12 volts, 2 x 15 volts 2 x 24, and 5 and 12 volts all wih easy configuration in series, parallel, symmetry.

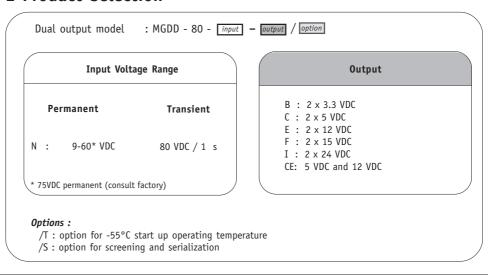
The total power is 80W with one single channel able to provide up to 75W in load unbalanced

mode.

All the modules are designed with LC network filters to minimize reflected input current ripple. The modules include a soft-start, an input undervoltage lock-out, a permanent short circuit and overload protection and an output overvoltage limitation to ensure efficient module protections. The soft-start allows current limitation and eliminates inrush current during start-up. The short circuit protection completely protects the modules against short-circuits of any duration by a shut-down and restores to normal when the overload is removed.

The modules are potted with a bi-component thermally conductive compound to ensure optimum power dissipation under harsh environmental conditions.

#### 2-Product Selection



REDEFINING THE SOURCE OF POWER



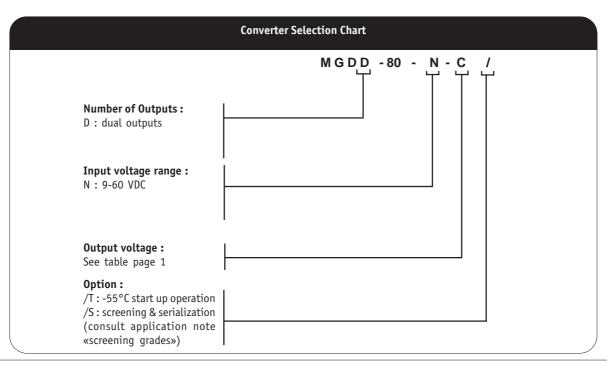


# 2- Product Selection (continued)

Input range	Output	Current per Output	Reference	Options
9-60 VDC	2 x 3.3 VDC	8 A	MGDD-80-N-B	/т, /s
9-60 VDC	2 x 5 VDC	8 A	MGDD-80-N-C	/т, /s
9-60 VDC	2 x 12 VDC	3.4 A	MGDD-80-N-E	/т, /s
9-60 VDC	2 x 15 VDC	2.7 A	MGDD-80-N-F	/т, /s
9-60 VDC	2 x 24 VDC	1.7 A	MGDD-80-N-I	/т, /s
9-60 VDC	5 VDC and 12 VDC	8A and 3.4A	MGDD-80-N-CE	/т, /s

Using various parallel or series connections of outputs, and the 80/110% trim capability (100/110% for 3.3 V output), allows to cover almost the complete range of output voltages from 3.3V to 52V as shown in the table below.

Reference	Parallel Connection	Series Connection	Symmetrical Connection
MGDD-80-N-B	3 - 3,6 VDC	6 - 7,2 VDC	+/-3 - +/-3,6 VDC
MGDD-80-N-C	4 - 5,5 VDC	8 - 11 VDC	+/-4 - +/-5,5 VDC
MGDD-80-N-E	9,6 - 13,2 VDC	19,2 - 26,4 VDC	+/-9,6 - +/-13,2 VDC
MGDD-80-N-F	12 - 16,5 VDC	24 - 33 VDC	+/-12 - +/-16,5 VDC
MGDD-80-N-I	19,2 - 26,4 VDC	38,4 - 52,8 VDC	+/-19,2 - +/-26,4 VDC
MGDD-80-N-CE	Forbiden	13.6-18.7 VDC	5/9.6V- 5.5/13.2 Vdc







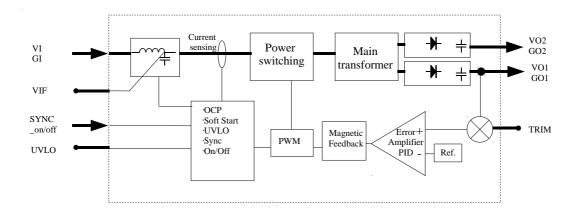
### 3- Block Diagram and Pin Description

The MGDD-80 is based on a new generation platform with a new topology allowing ultra wide input range up to 9 to 60 VDC input voltage.

The double output combined to the trim function allows multiple combinations to achieves most of the

common voltages i.e 3.3V-5V-9V-10V-12V-15V-24V-28V-30V-32V-48V-52V, and more over.

The MGDD-48 module block diagram and pin description are presented hereafter.



#### Pin Description:

**SYNC\_On/Off**: this pin is an input pin that allows user to synchronize to an external signal. See chapter synchronization for sync signal application. Connecting this pin to GI acts as an on/off function, shutting down the converter.

**UVLO:** this pin allows to user to increase the UVLO (Under voltage Lock out) threshold value, in order to stop converter operation when input voltage reaches a given value.

**OTP-D:** "OTP-disable mode" disbling the over temperature protection.

GI: input bus return lead.

**VIF:** this pin gives direct access to input filter capacitor that improves EMI performance.

VI: input positive bus lead.

TRIM: output voltage trimming input.

**G01:** return terminal of output number 1.

**V01:** output number 1 terminal. The output number 1 is considered as the main output, and must be mandatory loaded.

GO2: return terminal of output number 2.

**VO2:** output number 2 terminal.





# **4- Electrical Specifications**

Data are valid at +25°C, unless otherwise specified.

Parameter	Conditions	Limit or typical	Units	Dual Outputs MGDD-80 - N
Input				
Nominal input voltage	Full temperature range	Nominal	VDC	28
Permanent input	Full temperature range	M: M	VDC	0.60
voltage range (Ui)	Full load	Min Max.	VDC	9-60
Extended permanent input voltage range (Ui)	Consult factory Depends on reference	Min Max.	VDC	9-75
Transient input voltage	Full load	Minimum	VDC/s	80 / 1
Absolute max. input voltage	Power off	Maximum	VDC	100
Undervoltage lock-out (UVLO)	Turn-on voltage	Nominal	VDC	9
Pin UVLO not connected	Hysteresis	Nominal	VDC	1
Undervoltage lock-out range	Trim range	Nominal	VDC	9- Vin max.
Start up time on power-up	Ui nominal, full load resistive	Maximum	ms	30
Start up time on SD release	Ui nominal, full load resistive	Maximum	ms	30
Reflected ripple current	Ui nominal, full load at switching freq. BW = 20MHz 1µF on Vif, 33µF on Vin	Maximum	% Inom.	10
Input power in inhibit mode	Ui nominal inhibit mode	Nominal	W	0.015
Input power in short circtui	Ui nominal, Vo shorted	Nominal	W	12
Output				
		Nominal	VDC	2 x 3.3
		Nominal	VDC	2 x 5
Output voltage		Nominal	VDC	2 x 12
		Nominal	VDC	2 x 15
		Nominal	VDC	2 x 24
Set Point accuracy	Ambient temperature : +25°c Ui nominal, 75% load	Maximum	%	+/- 2
Total output power with both outputs loaded	Full temperature range Ui min. to max.	Maximum	W	80 (53W for 3.3V output)
Maximum power per channel	Full temperature range Ui min. to max.	Maximum	W	70 (45W for 3.3V output)
Output current per output		Nominal	Λ.	
5V & 3.3Voutput	Full temperature range	Nominal	A	8
12V output	Full load see note (1)		A	3.4
15V output	Ui min. to max.	Nominal	A	2.7
24V output		Nominal	Α	1.7
Ripple output voltage **				
5V&3.3V output	Ui nominal	Maximum	mVpp	150
12V output	Full load	Maximum	mVpp	300
15V output	BW = 20MHz	Maximum	mVpp	300
24V output	20.11.12	Maximum	mVpp	500
Output regulation *	Ui min. to max.			
(Line + load + thermal)	0% to full load	Maximum	%	+/- 2
Cross load output regulation	Ui min. to max. V01 at nominal load V02 from 25% to full load As Function of nominal set point value	Maximum	%	+12/-0
Output voltage trim	As a function of output voltage	Minimum Maximum	% %	80 110
Maximum admissible capacitive load		-		<u> </u>
5V & 3.3V ouput	Ui nominal	Maximum	μF	1 000
12V output	Full load	Maximum	μF	470
15V output	Per output	Maximum	μι μF	330
24V ouput	rei output	Maximum	μr μF	220
Efficiency	Ui nominal Full load	Typical	%	up to 92
	I MIL IVUU			

Note \*: Regulation is measured with both outputs in parallel configuration.

Note \*\*: The ripple output voltage is the periodic AC component imposed on the output voltage, an aperiodic and random component (noise) has also to be considered.

This noise can be reduced by adding 1 external decoupling capacitor connected between Gin and Gout. These capacitance should be layed-out as close as possible from the converter. The ripple output voltage is measured by connecting a ceramic chip capacitor Co accross Vo and Go pins (C=100µF if Vo>5Vdc C=10µF if Vo>5Vdc)





# **4- Electrical Specifications**

Data are valid at +25°C, unless otherwise specified.

Parameter	Conditions	Limit or typical	Units	Dual Outputs MGDD-80 - N-CE
Input				
Nominal input voltage	Full temperature range	Nominal	VDC	28
Permanent input voltage range (Ui)	Full temperature range Full load	Min Max.	VDC	9-60
Extended permanent input voltage range (Ui)	Consult factory Depends on reference	Min Max.	VDC	9-75
Transient input voltage	Full load	Minimum	VDC/s	80 / 1
Absolute max. input voltage	Power off	Maximum	VDC	100
Undervoltage lock-out (UVLO)	Turn-on voltage	Nominal	VDC	9
Pin UVLO not connected	Hysteresis	Nominal	VDC	1
Undervoltage lock-out range	Trim range	Nominal	VDC	9- Vin max.
Start up time on power-up	Ui nominal, full load resistive	Maximum	ms	30
Start up time on SD release	Ui nominal, full load resistive	Maximum	ms	30
<u> </u>	Ui nominal, full load at			
Reflected ripple current	switching freq. BW = 20MHz	Maximum	%	10
pp	1μF on Vif, 33μF on Vin		Inom.	
Input power in inhibit mode	Ui nominal	Nominal	W	0.015
	inhibit mode	<b>-</b> • •		
no load input power	Ui nominal	Typical	W	0.5
Input power in short circtui	Ui nominal, Vo shorted	Nominal	W	12
Output				
Output voltage		Nominal	VDC	5 and 12
Set Point accuracy Vo1: 5V output	Ambient temperature : +25°c Ui nominal, 75% load	Maximum	%	+/- 2
Set Point accuracy Vo2:12V output	Ambient temperature : +25°c Ui nominal, 75% load	Maximum	%	+/- 5
Total output power with both outputs loaded	Full temperature range Ui min. to max.	Maximum	W	80
Maximum power per channel	Full temperature range Ui min. to max.	Maximum	W	70
Output current per output	Full temperature range			
Vo1: 5V output	Full load see note (1)	Nominal	A	8
Vo2:12V output	Ui min. to max.	Nominal	Α	3.4
Ripple output voltage **	Ui nominal	Mandar	mal for :-	450
Vo1: 5V output Vo2: 12V output	Full load BW = 20MHz	Maximum Maximum	mVpp mVpp	150 300
Output regulation *	Ui min. to max.			
(Line + load + thermal)	0% to full load	Maximum	%	+/- 2
Cross load output regulation	Ui min. to max. V01 at nominal load V02 from 25% to full load	Maximum	%	+8/-6
Output voltage trim	As a function of output voltage	Minimum Maximum	%	80 110
Maximum admissible capacitive load	Ui nominal			
5V ouput	Full load	Maximum	μF	1 000
12V output	Per output	Maximum	μF	470
Efficiency	Ui nominal Full load	Typical	%	up to 89

Note \*: Regulation is measured with both outputs in parallel configuration.

Note \*\*: The ripple output voltage is the periodic AC component imposed on the output voltage, an aperiodic and random component (noise) has also to be considered. This noise can be reduced by adding 1 external decoupling capacitor connected between Gin and Gout. These capacitance should be layed-out as close as possible from the converter. The ripple output voltage is measured by connecting a ceramic chip capacitor Co accross Vo and Go pins (C=100µF if Vo<5Vdc C=10µF if Vo>5Vdc)

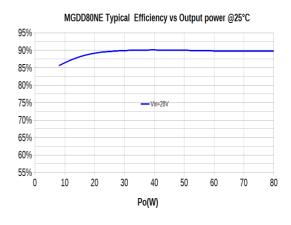


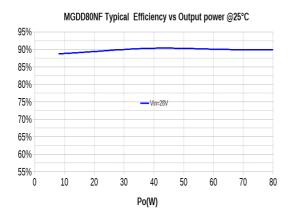


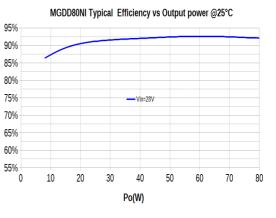
# 4- Electrical Specifications (continued)

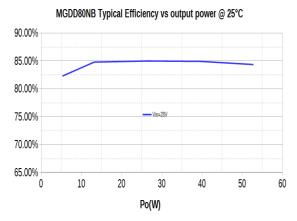
Data are valid at +25°C, unless otherwise specified.

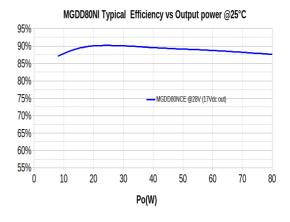
















# 5- Switching Frequency

Characteristics	Conditions	Limit or typical	Specification
Switching frequency	Full temperature range Ui min. to max. No load to full load	Nominal, fixed	330 KHz
Synchronization frequency range	Frequency of external synchronization signal	Minimum Maximum	270 KHz 360 KHz

# 6- Isolation

Parameters	Conditions	Limit or typical	Specifications
Isolation voltage (Case not connected)	Input to output Between outputs	Minimum Minimum	1 500 Vdc/ 1 min 300 Vdc
Isolation safety rating	Input to output Between outputs	/	Functionnal
Isolation capacitance	Input to input	Typical	1 nF
Isolation resistance	Input to case 500 Vdc Output to case 500 Vdc	Minimum Minimum	100 M0hm 100 M0hm

# **7- Protection Functions**

Characteristics	Protection Device	Recovery	Limit or typical	Specifications
Input undervoltage lock-out (UVL0)	Turn-on, turn-off circuit with hysteresis cycle	Automatic recovery	Turn-on nominal Turn-off nominal	See section 3
Output current limitation protection (OCP)	Straight line current limitation	Automatic recovery	Typical Minimum	160% of Inom. 105% of Inom.
Output overvoltage protection (OVP)	Overvoltage protection with latch-up	Automatic recovery	Typical	130% (160% for MGDD80NB) of output voltage
Over temperature protection (OTP)	Thermal device with hysteresis cycle	Automatic recovery	Nominal Hysteresis	125°C 10°C

# 8- Reliability Data

Characteristics	Conditions	Temperature	Specifications
	Ground benign (Gb)	Case at 40°C	2 253 000 Hrs
Mean Time Between Failure (MTBF) According to MIL-HDBK-217F	Ground fixed (Gf)	Case at 40°C Case at 70°C	852 000 Hrs 398 000 Hrs
	Airborne, Inhabited Cargo (AIC)	Case at 40°C Case at 70°C	489 000 Hrs 241 4000 Hrs
Mean Time Between Failure (MTBF) According to IEC-62380-TR	Aircraft Civilian	Ambient at 25°C 100% time on	Consult factory





## 9- Electromagnetic Interference and Surge

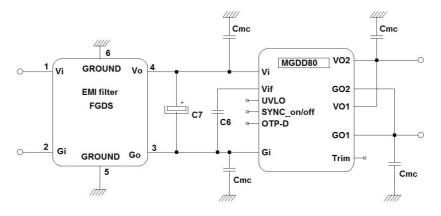
Electromagnetic Interference requirements according to MIL-STD-461C/D/E/F standards can be easily achieved as indicated in the following section. The following table resumes the different sections covered by these standards.

Standard Requirements	MIL-STD-461C Standard	MIL-STD-461D/E/F Standard	Compliance with GAIA Converter Module & common mode capacitance
Conducted emission (CE):	CF 01	CF 404	
Low frequency High frequency	CE 03	CE 101 CE 102	compliant module stand-alone compliant with additionnal filter
Conducted susceptibility (CS):			
Low frequency High frequency	CS 01 CS 02	CS 101 CS114	compliant with additionnal filter compliant with additionnal filter
Radiated emission (RE):			
Magnetic field	RE 01	RE 101	compliant module stand-alone
Electrical field	RE 02	RE 102	compliant module stand-alone
Radiated susceptibility (RS):			
Magnetic field	RS 01	RS 101	compliant module stand-alone
Electrical field	RS 03	RS 103	compliant module stand-alone

#### 9-1 Module Compliance with MIL-STD-461 Standards

To meet the latest US military standards MIL-STD-461 requirements and in particular the conducted noise emission CE102 (and also CE03) requirements, GAIA Converter can propose an EMI filter module. In addition common mode capacitances  $C_{mc}$  (10nF/rated voltage depending on isolation requirement) connected between power pins and chassis or ground powerplane need to be implemented. When output channels are intended to be connected to load through long wire, it could be necessary to use additionnaly common mode inductors on each outputs. For a use at 28Vdc C7 can be rated to  $100\mu F$ .

Please consult FGDS series datasheets for further details.







### 10- Thermal Characteristics

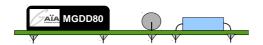
Characteristics	Conditions	Limit or typical	Performances
Operating ambient temperature range	Ambient temperature	Minimum Maximum	- 40°C see below
Operating case temperature range	Case temperature	Minimum Maximum	- 40°C 105°C
Storage temperature range	Non functionning	Minimum Maximum	- 55°C + 125°C

To calculate the maximum ambient temperature at which the converter will be able to operate, the following parameters are required:

- Tcase = maximum case temperature the converter can operate
- Tmax = maximum ambient temperature the converter can operate
- Pout = effective output power used (see also page therafter curves of maximum power)
- Rth(c) = thermal resistance case to ambient of the converter (see table below) in free air natural convection
- Rth(tot) = thermal resistance of converter and its heatsink (if used)
- Rth(hs) = Thermal resistance of heat sink (if used)
- $\eta$  = converter efficiency

There are 3 mounting possibilities as described below:

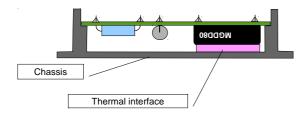
#### **Converter Stand-Alone Mounting**



To calculate the maximum ambient temperature at which the converter can operate, the following formula can be applied:

Tmax = Tcase - Rth(c) x Pout( $1/\eta - 1$ )

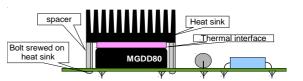
#### **Converter with Chassis Mounting**



If the thermal interface resistance can be neglected, the maximum chassis temperature is equal to the maximum converter case temperature.

Tmax chassis = Tmax.

#### **Converter with Heatsink Mounting**



To calculate the maximum ambient temperature the converter can operate, the following formula can be applied :

Tmax = Tcase - Rth(tot) x Pout( $1/\eta - 1$ )

The most sensitive parameter in this formula is the Rth(tot) value.

Rth(tot) depends on the thermal resistance of the converter Rth(c) in the mounting configuration and the thermal resistance of the heatsink Rth(hs).

- The Rth(c) depends on ambient temperature, the way the converter is tied to the PCB, position,PCB copper track and power plane length.
  - Also, in general Rth(c) decreases as temperature is increases.
- Rth(hs): Rth(hs) value is highly depending on how the heatsink is connected to case.

The value of Rth(tot) can be evaluated with the below definition:

Rth(c)\*Rth(hs)/(Rth(c)+Rth(hs)) < Rth(tot) < Rth(c)

The table hereafter gives some example of thermal resistance for different heat transfert configurations.

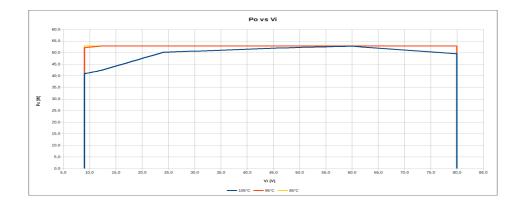
Heat transfert	Thermal resistance heatsink to air Rth(h-a)	Conditions	Global resistance
	Rth(c) : no Heatsink baseplate only	Ambient 60°C, converter on PCB	9 °C/W
Free air cooling only			

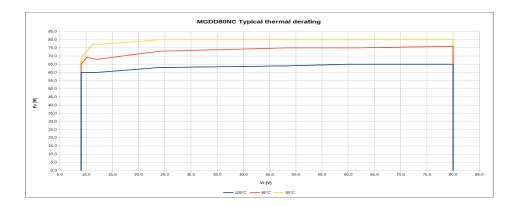


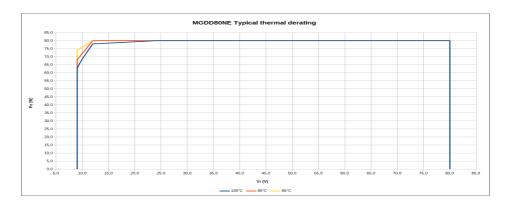


# 10- Thermal Characteristics (continued)

A maximum power for a Tcase of 105°C depending on input voltage is given in the following curves



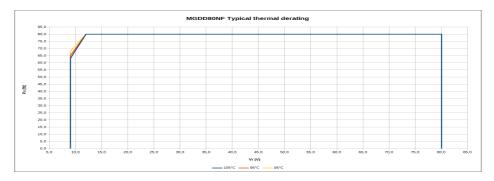


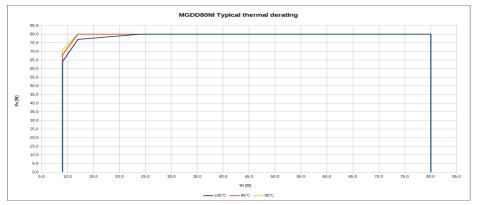






# 10- Thermal Characteristics (continued) A maximum power for a Tcase of 105°C depending on input voltage is given in the following curves









# 11- Environmental Qualifications

The modules have been subjected to the following environmental qualifications.

Characteristics	Conditions	Severity	Test procedure
Climatic Qualificati	ons		
Life at high temperature	Duration Temperature / status of unit	Test D: 1 000 Hrs @ 105°C case, unit operating @ 125°C ambient, unit not operating	MIL-STD-202G Method 108A
Altitude	Altitude level C Duration Climb up Stabilization Status of unit	40 000 ft@-55°C 30 min. 1 000 ft/min to 70 000 ft@-55°C, 30 min. unit operating	MIL-STD-810E Method 500.3
Humidity cyclic	Number of cycle Cycle duration Relative humidity variation Temperature variation Status of unit	10 Cycle I: 24 Hrs 60 % to 88 % 31°C to 41°C unit not operating	MIL-STD-810E Method 507.3
Humidity steady	Damp heat Temperature Duration Status of unit	93 % relative humidity 40°C 56 days unit not operating	MIL-STD-202G Method 103B
Salt atmosphere	Temperature Concentration NaCl Duration Status of unit	35°C 5 % 48 Hrs unit not operating	MIL-STD-810E Method 509.3
Temperature cycling	Number of cycles Temperature change Transfert time Steady state time Status of unit	200 -40°C / +85°C 40 min. 20 min. unit operating	MIL-STD-202A Method 102A
Temperature shock	Number of shocks Temperature change Transfert time Steady state time Status of unit	100 -55°C / +105°C 10 sec. 20 min. unit not operating	MIL-STD-202G Method 107G
Mechanical Qualific	cations		
Vibration (Sinusoidal)	Number of cycles Frequency / amplitude Frequency / acceleration Duration Status of unit	10 cycles in each axis 10 to 60 Hz / 0.7 mm 60 to 2 000 Hz / 10 g 2h 30 min. per axis unit not operating	MIL-STD-810D Method 514.3
Shock (Half sinus)	Number of shocks Peak acceleration Duration Shock form Status of unit	3 shocks in each axis 100 g 6 ms 1/2 sinusoidal unit not operating	MIL-STD-810D Method 516.3
Bump (Half sinus)	Number of bumps Peak acceleration Duration Status of unit	2 000 Bumps in each axis 40 g 6 ms unit not operating	MIL-STD-810D Method 516.3





### 12- Description of Protections

The MGDD-80 series includes 2 types of protection devices.

#### 12-1 Input Undervoltage Lockout (UVLO)

An input undervoltage protection will inhibit the module when input voltage drops below the lock-out turn-off threshold (see section 3 for value) and restores to normal operation automatically when the input voltage rises above the lock-out turn-on threshold.

The UVLO voltage can be adjusted using an external resistor (Ruvlo) connected between pin 2 and Gi. This value can be adjusted in order to allow converter to shut down properly depending on the input bus (or battery) voltage value. Ruvlo can be determined using the following formula

Ruvlo (K
$$\Omega$$
) =  $\frac{110}{\text{Vuvlo} - 8.51}$  - 1

Ruvlo = trimming resistance

Vuvlo = desired turn-on voltage

Values are in KOhms

#### 12-2 Output Over Current Protection (OCP)

The MGDD-80 Series features an over-current protection circuit that detects short circuit or over current and protects the module according to the hiccup graph .

The maximum detection current Id is depending on input voltage Vin and temperature. When OCP is triggered, the converter falls into hiccup mode, testing periodically if the overload is still present. The module restart automatically in soft-start to normal operation when overcurrent is removed. Td (detection time) and Th (hiccup period) are depending on Vin and temperature.

#### 12-3 Output Overvoltage Protection (OVP)

The MGDD-80 Series features an ouput overvoltage protection circuit that will shut down the converter if the voltage at the output is higher than a threshold (see section 7) of the nominal output voltage to protect the output circuitry from damage.

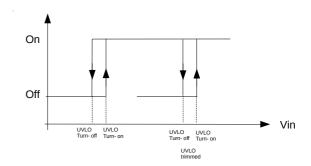
The converter will automatically restart in soft-start mode when the overvoltage is removed.

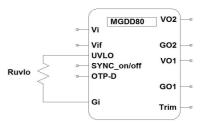
#### 12-4 Over Temperature Protection (OTP)

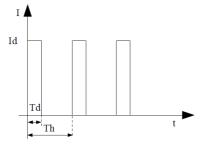
A thermal protection device will inhibit the module when internal temperature reaches the specified value (see section 7) as long as the overheat is present and restores to normal operation automatically when overheat is removed. The efficiency of the OTP function is warranty with the module mounted on a heatsink.

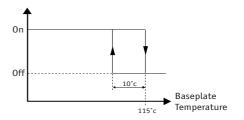
#### 12-5 OTP-Disable (OTP-D)

The MGDD80 features a OTP-Disable mode. When the OTP-D pin is pulled down, the overtemperature protection is disabled to allow the converter to operate above its maximum case temperature. This mode of operation should only be used for limited durations to avoid damaging the converter prematurely. Once the pin is released, the OTP function is enabled again.













# 13- Description of Functions

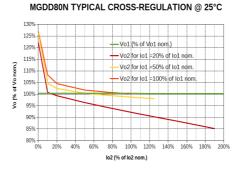
#### 13-1 Connection of Outputs

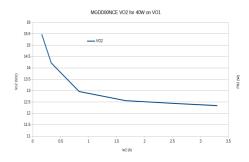
The outputs of MGDD-80 can be connected in various configurations such as:

- connections in series
- connection in parallel (non alowed for MGDD80-N-CE)
- connection in symmetry
- connection in independance

Please note that regulation is achieved through output V01/G01 referenced as primary output. When connected in symmetry or independant configurations with unbalanced loads, V01/G01 has to be loaded at 4W minimum to insure proper operation of the converter.

VO2/GO2 limits: The VO2/GO2 output referenced as secondary output may stay unloaded but in that case its regulation may drift-up. VO2 drift increases with VO1 load and can reach up to 130% (typical) of VO2 nominal voltage in worst case. A 10% load on VO2 will bring back the drift within lower values as per graph.

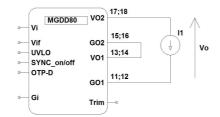




Vo2 regulation versus current Io2 as % of nominal Io2nom @ Vo1 various fixed load as a % of nominal current Io1nom

#### 13-1-1 Connection of Outputs in Series

Outputs connected in series allow to achieve 10V,24V, 30V or 48V output voltages up to 80W total power. These values can be extended using trim adjustment.



### 13-1-3 Connection of Outputs in Symmetry

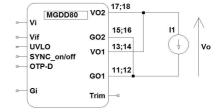
Outputs connected in symmetry allow to achieve +/-5V, +/-12V, +/-15V or +/-24V voltages (+/-40W on each channel) with possible unbalanced loadup to 70W on output 1, 10W on output 2 and vice versa.

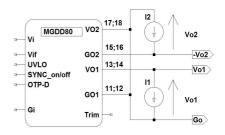
#### 13-1-4 Connection of Outputs in Independance

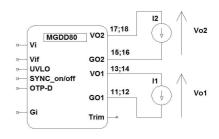
Outputs connected independantly with floating voltage between each other can be achieved for 2x5V, 2x12V, 2x15V or 2x24V voltages (40W each) with possible unbalanced load up to 70W on output 1, 10W on output 2 and vice versa.

### 13-1-2 Connection of Outputs in Parallel

Outputs connected in parallel allow to achieve single output 5V, 12V, 15V or 24V up to 80W power. These values can be extended using trim adjustment.







# 4



### 13- Description of Functions (continued)

#### 13-2 Trim Function

The output voltage Vo1 may be trimmed in a range of 80% to 110% of the nominal output voltage (100%/110% for 3,3 Vdc output voltage) via a single external trimpot or fixed resistor.

The VO2 output will be automatically trimmed to the same value as VO2, whatever the outputs combination is.

#### **Trim Up Function**

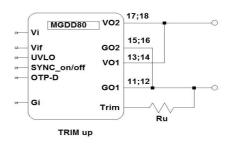
Do not attempt to trim the module higher than 110% of nominal output voltage as the overvoltage protection may trigger.

Also do not exceed the maximum rated output power when the module is trimmed up.

The trim up resistance must be calculated with the following formula :

$$R_U(k \Omega) = 4.7 \cdot \frac{VO_{nom} - 1.225}{VO - VO_{nom}} - 0.27$$

Ru is the trim resistor value in K0hm  $V0_{nom}$  is the nominal output voltage 1 V0 is the desired trimmed output voltage



#### **Trim Down Function**

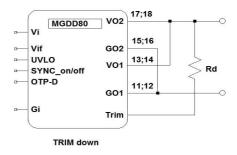
Do not trim down more than -20% of nominal output voltage otherwise the module may be damaged.

The available output power is reduced by the same percentage that the output voltage is trimmed down.

The trim down resistance must be calculated with the following formula :

$$R_D(k\Omega) = 4.7 \cdot (\frac{VO_{nom}}{1,225} - 1) \cdot \frac{VO - 1,225}{VO_{nom} - VO} - 0.27$$

Rd is the trim resistor value in K0hm  $\rm VO_{nom}$  is the nominal output voltage 1 V0 is the desired trimmed output voltage







# 13- Description of Functions (continued)

#### 13-3 On/Off (SYNC\_On/Off) Function

The control pin 1 (SYNC\_On/Off) can be used for applications requiring On/Off operation. This may be done with an open collector transistor, a switch, a relay or an optocoupler. Several converters may be disabled with a single switch by connecting all SYNC\_On/Off pins together.

- The converter is disabled by pulling low pin 1.
- No connection or high impedance on pin 1 enables the converter.

By releasing the On/Off function, the converter will restart within the start up time specifications given in table section 3

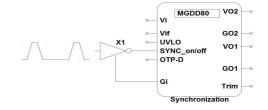
	□ Vi MGDD8	00 VO2
	□— Vif	GO2
	UVLO SYNC_on/of	vo1
	□ OTP-D	
M1		GO1 —
—— <u> </u>	Gi	Trim
	ON/OFF	control

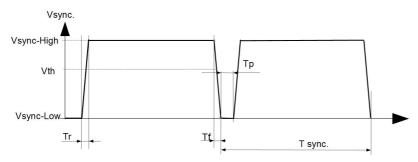
Parameter	Unit	Min.	Тур.	Max.	Notes, conditions
On/Off module enable voltage	Vdc	0.7	/	3.3	Open, the switch must not sink more than 50µA
On/Off module disable voltage	Vdc	/	/	0.5	The switch must be able to sink 0,5mA
On/Off module enable delay	ms	/	/	30	The module restarts with the same delay after alarm mode removed
On/Off module disable delay	μs	/	/	100	Vi nominal, full load

#### 13-4 Synchronization (SYNC\_On/Off) Function

The MGDD-80 voltage series provides an external synchronization function through the SYNC-On/Off pin. SYNC-On/Off pin is an <u>input only</u> and is referenced to Gi. Automatic synchronization of multiple units (all Sync pins connected) is not possible.

This pin can be driven directly by using a LV TTL (3,3V) gate. SYNC\_On/Off pin is internally pulled up to 3V (logic level 1). It is possible to synchronize the module by using an open collector gate (transistor, optocoupler, ...). Minimum 3,3V LV TTL rise time (tr) and fall time (tf) are 20ns. The module can lock on frequency above or below its free-run frequency 2.52  $\mu$ s < Tsync < 3.78 $\mu$ s and 0.3  $\mu$ s<Tp< 2.7 $\mu$ s.

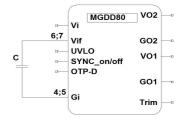




#### 13-5 Input Filter Compensation (VIF)

In most applications the internal filter capacitance value is sufficient to insure stability of the MGDD80.

However for stringent application (low input voltage) it is possible to improve the converter stability and to reduce the input current ripple for better EMI performance, by adding a capacitor accross "VIF" pin and "Gin" pin. «VIF» pin is a direct access to the capacitor of the internal LC input filter.

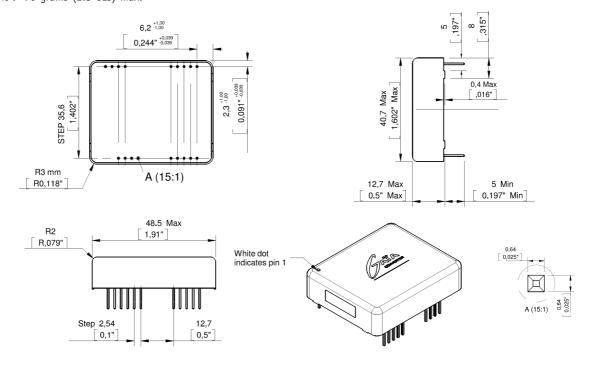






#### 14- Dimensions

Dimension are given in mm. Tolerance: +/- 0,2 mm (+/- 0.01 ") unless otherwise indicated. All dimensions specified "Min" or "Max" are subjected to tolerance Min $^{+0.5/-0mm}$  and Max $^{+0/-0.5mm}$ . Weight: 70 grams (2.5 0zs) max.



#### 15- Materials

Case: Metallic black anodized coating.
Pins: Flash gold plating over nickel underplate.

## 16- Product Marking

Upper face : Company logo.

Side face: Module reference, option, date code: year and week of manufacturing.

#### 17- Connections

The MGDD-80 series has been designed for on-board mounting. it is recommended not to lay-out any component under the module.



Bottom view

Pin	Dual			
1	Sync-SD			
2	UVLO			
3	OTP-D			
4, 5	- Input (Gi)			
6, 7	VIF			
8, 9	+ Input (Vi)			
10	0 Vtrim			
11, 12	- Output 1 (Go1)			
13, 14	+ Output 1 (Vo1)			
15, 16	- Output 2 (Go2)			
17, 18	18 + Output 2 (Vo2)			





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